



SBOS414B-AUGUST 2007-REVISED DECEMBER 2008

# 1.8V, 700nA, Zerø-Crossover RAIL-TO-RAIL I/O OPERATIONAL AMPLIFIER

#### **FEATURES**

nanoPOWER:

OPA369: 800nAOPA2369: 700nA/ch.

LOW OFFSET VOLTAGE: 250μV

ZERO-CROSSOVER

LOW OFFSET DRIFT: 0.4μV/°C

DC PRECISION:
 CMRR: 114dB
 PSRR:106dB

AOL: 134dB

GAIN-BANDWIDTH PRODUCT: 12kHz

SUPPLY VOLTAGE: 1.8V to 5.5V

microSIZE PACKAGES:

SC70-5, SOT23-5, MSOP-8

#### **APPLICATIONS**

- BATTERY-POWERED INSTRUMENTS
- PORTABLE DEVICES
- MEDICAL INSTRUMENTS
- TEST EQUIPMENT
- LOW-POWER SENSOR SIGNAL CONDITIONING

#### DESCRIPTION

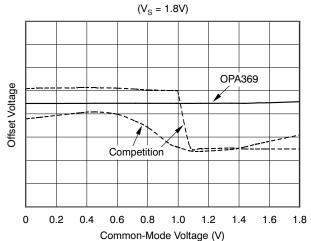
The OPA369 and OPA2369 are ultra-low-power, low-voltage operational amplifiers from Texas Instruments designed especially for battery-powered applications.

The OPAx369 operates on a supply voltage as low as 1.8V and has true rail-to-rail operation that makes it useful for a wide range of applications. The zerø-crossover feature resolves the problem of input crossover distortion that becomes very prominent in low voltage (< 3V), rail-to-rail input applications.

In addition to *micro*size packages and very low quiescent current, the OPAx369 features 12kHz bandwidth, low offset drift (1.75 $\mu$ V/°C, max), and low noise 3.6 $\mu$ V<sub>PP</sub> (0.1Hz to 10Hz).

The OPA369 (single version) is offered in an SC70-5 package. The OPA2369 (dual version) comes in both MSOP-8 and SOT23-8 packages.

# OFFSET VOLTAGE vs COMMON-MODE VOLTAGE



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### **ABSOLUTE MAXIMUM RATINGS**(1)

Over operating free-air temperature range (unless otherwise noted).

			VALUE	UNIT
Supply Voltage		$V_{S} = (V+) - (V-)$	+7	V
Single Input	Voltage <sup>(2)</sup>		(V-) -0.5 to (V+) + 0.5	V
Terminals	Current <sup>(2)</sup>		±10	mA
Output Short-0	tput Short-Circuit <sup>(3)</sup>		Continuous	
Ambient Oper	Ambient Operating Temperature		-55 to +125	°C
Ambient Stora	ge Temperature		-65 to +150	°C
Junction Temp	perature	T <sub>J</sub>	+150	°C
	Human Body Model	(HBM)	4000	V
ESD Ratings	Charged Device Model	(CDM)	1000	V
	Machine Model	(MM)	200	V

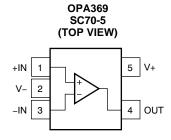
- (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not supported.
- (2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current limited to 10mA or less.
- (3) Short-circuit to V<sub>S</sub>/2, one amplifier per package.

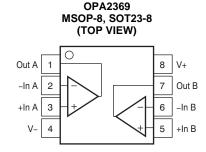
#### PACKAGE/ORDERING INFORMATION(1)

PRODUCT	PACKAGE-LEAD	PACKAGE DESIGNATOR	PACKAGE MARKING
OPA369	SC70-5	DCK	CJS
OPA2369	MSOP-8	DGK	OCCQ
	SOT23-8	DCN	OCBQ

(1) For the most current package and ordering information see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

#### **PIN CONFIGURATIONS**







# ELECTRICAL CHARACTERISTICS: $V_S = +1.8V$ to +5.5V

**BOLDFACE** limits apply over the specified temperature range,  $T_A = -40^{\circ}C$  to +85°C. At  $T_A = +25^{\circ}C$ , and  $R_L = 100k\Omega$  connected to  $V_S/2$ , unless otherwise noted.

			_			
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT	
OFFSET VOLTAGE						
Input Offset Voltage V <sub>OS</sub>			250	750	μV	
over Temperature				1	mV	
Drift dV <sub>OS</sub> /dT			0.4	1.75	μ <b>V/°C</b>	
vs Power Supply PSRR	$V_S = 1.8V \text{ to } 5.5V$		5	20	μ <b>V/V</b>	
Channel Separation	dc		140		dB	
	f = 1kHz		120		dB	
INPUT VOLTAGE RANGE						
Common-Mode Voltage Range V <sub>CM</sub>		(V-)		(V+)	V	
Common-Mode Rejection Ratio CMRR	$(V-) \le V_{CM} \le (V+)$	100	114		dB	
over Temperature	$(V-) \le V_{CM} \le (V+)$	90			dB	
INPUT BIAS CURRENT						
Input Bias Current I <sub>B</sub>			10	50	pA	
over Temperature			See Figure 16		pА	
Input Offset Current I <sub>OS</sub>			10	50	pA	
INPUT IMPEDANCE						
Differential			10 <sup>13</sup>    3		Ω    pF	
Common-Mode			10 <sup>13</sup>    6		Ω    pF	
NOISE			- 11 -			
Input Voltage Noise	f = 0.1Hz to 10Hz		3.6		$\mu V_{PP}$	
Input Voltage Noise Density	f = 100Hz		220		nV/√Hz	
, , , , , , , , , , , , , , , , , , , ,	f = 1kHz		290		nV/√Hz	
Current Noise Density	f = 1kHz		1		fA/√Hz	
OPEN-LOOP GAIN						
Open-Loop Voltage Gain A <sub>OL</sub>	$100 \text{mV} \le \text{V}_{\text{O}} \le (\text{V+}) - 100 \text{mV},$	444	404		-ID	
	$R_L = 100k\Omega$	114	134		dB	
Over Temperature	100mV $\leq$ V <sub>O</sub> $\leq$ (V+)-100mV, R <sub>L</sub> = 100kΩ	100			dB	
	$500\text{mV} \le V_{\text{O}} \le (\text{V+})-500\text{mV},$ $R_{\text{L}} = 10\text{k}\Omega$	114	134		dB	
Over Temperature	500mV $\leq$ V <sub>O</sub> $\leq$ (V+)-500mV, R <sub>L</sub> = 10k $\Omega$	90			dB	
ОИТРИТ						
Voltage Output Swing from Rail	$R_L = 100k\Omega$			10	mV	
	$R_L = 10k\Omega$			25	mV	
Short-Circuit Current I <sub>SC</sub>			10		mA	
Capacitive Load Drive C <sub>LOAD</sub>			See Figure 20		pF	
FREQUENCY RESPONSE						
Gain-Bandwidth Product GBW			12		kHz	
Slew Rate SR	G = +1		0.005		V/µs	
Overload Recovery Time	V <sub>IN</sub> × Gain > V <sub>S</sub>		250		μs	



# ELECTRICAL CHARACTERISTICS: $V_s = +1.8V$ to +5.5V (continued)

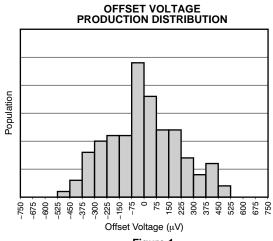
**BOLDFACE** limits apply over the specified temperature range,  $T_A = -40^{\circ}C$  to +85°C. At  $T_A = +25^{\circ}C$ , and  $R_L = 100k\Omega$  connected to  $V_S/2$ , unless otherwise noted.

		C	OPA369, OPA2369				
PARAMETER	CONDITIONS	MIN	TYP MAX		UNIT		
POWER SUPPLY							
Specified Voltage	s	1.8		5.5	V		
Quiescent Current	I <sub>OUT</sub> = 0A						
OPA369			0.8	1.2	μΑ		
OPA2369 (per channel)			0.7	1	μΑ		
Over Temperature							
OPA369				1.45	μ <b>Α</b>		
OPA2369 (per channel)				1.25	μ <b>Α</b>		
TEMPERATURE RANGE							
Specified Range	T <sub>A</sub>	-40		+85	°C		
Operating Range	T <sub>A</sub>	-55		+125	°C		
Thermal Resistance $\theta$	A						
SC70			250		°C/W		
SOT23			223		°C/W		
MSOP			252		°C/W		

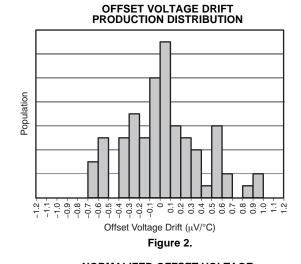


#### TYPICAL CHARACTERISTICS

At  $T_A = +25$ °C,  $V_S = 5$ V, and  $R_L = 100$ k $\Omega$  connected to  $V_S/2$ , unless otherwise noted.







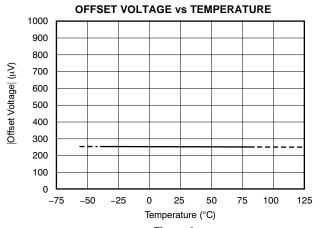


Figure 3.

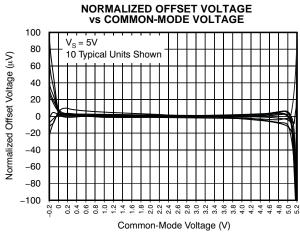


Figure 4.

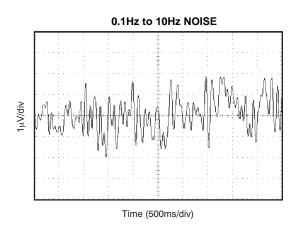
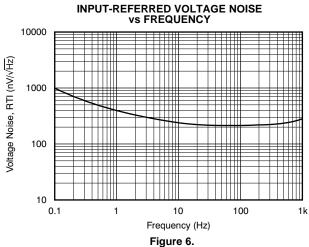


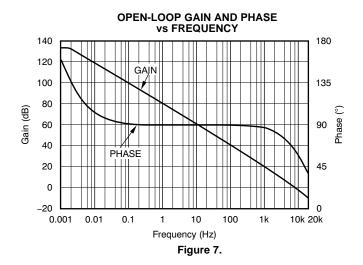
Figure 5.

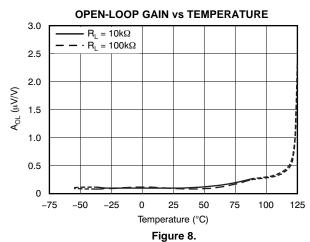


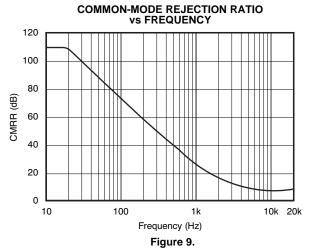


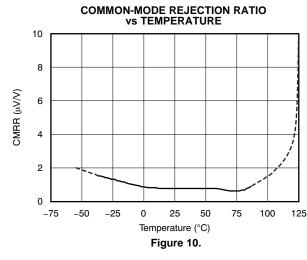
### **TYPICAL CHARACTERISTICS (continued)**

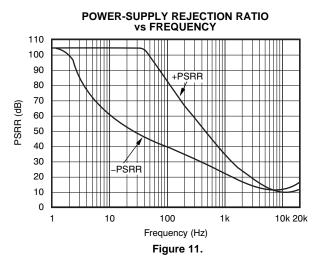
At  $T_A$  = +25°C,  $V_S$  = 5V, and  $R_L$  = 100k $\Omega$  connected to  $V_S/2$ , unless otherwise noted.

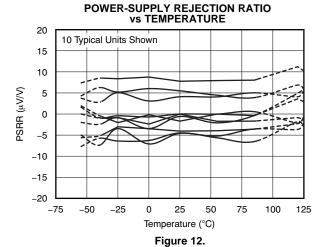








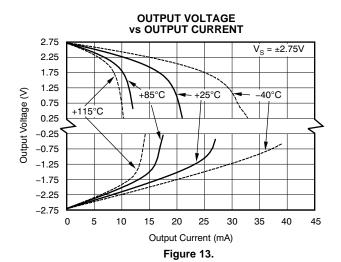






### **TYPICAL CHARACTERISTICS (continued)**

At  $T_A$  = +25°C,  $V_S$  = 5V, and  $R_L$  = 100k $\Omega$  connected to  $V_S/2$ , unless otherwise noted.

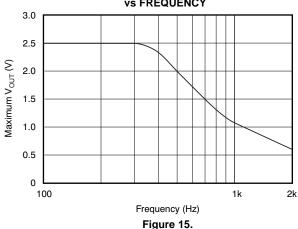


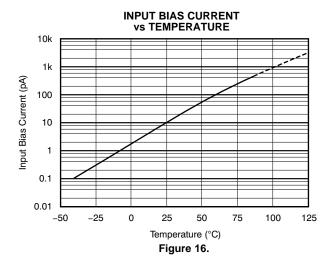
vs TEMPERATURE 25 Output Voltage Swing-from-Rail (mV) 20  $R_L = 10k\Omega$ 15 10 5  $R_1 = 100k\Omega$ 0  $R_L = 100k\Omega$ -5 -10  $R_L = 10k\Omega$ -15 -20 -25 25 50 125 -75 -50 -25 100

**OUTPUT VOLTAGE SWING-FROM-RAIL** 

Temperature (°C) Figure 14.







### QUIESCENT CURRENT vs TEMPERATURE

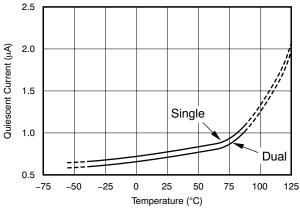


Figure 17.

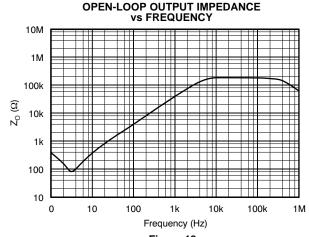
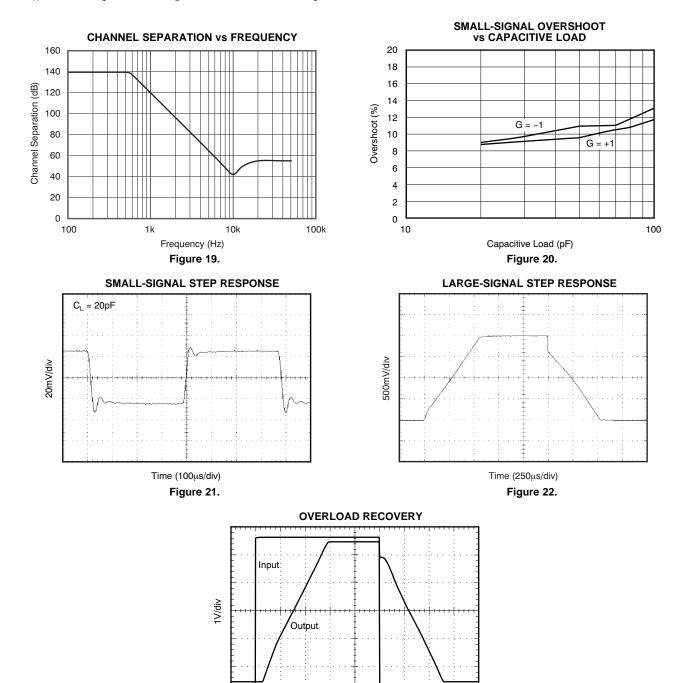


Figure 18.



### **TYPICAL CHARACTERISTICS (continued)**

At  $T_A$  = +25°C,  $V_S$  = 5V, and  $R_L$  = 100k $\Omega$  connected to  $V_S/2$ , unless otherwise noted.



Time (500 $\mu$ s/div) Figure 23.



#### **APPLICATION INFORMATION**

The OPA369 family of operational amplifiers minimizes power consumption and operates on supply voltages as low as 1.8V. Power-supply rejection ratio (PSRR), common-mode rejection ratio (CMRR), and open-loop gain ( $A_{OL}$ ) typical values are in the range of 100dB or better.

When designing for ultralow power, choose system components carefully. To minimize current consumption, select large-value resistors. Any resistors will react with stray capacitance in the circuit and the input capacitance of the operational amplifier. These parasitic RC combinations can affect the stability of the overall system. A feedback capacitor may be required to assure stability and limit overshoot or gain peaking.

Good layout practice mandates the use of a  $0.1 \mu F$  bypass capacitor placed closely across the supply pins.

#### **OPERATING VOLTAGE**

OPA369 series op amps are fully specified and tested from +1.8V to +5.5V (±0.9V to ±2.75V). Parameters that vary significantly with supply voltage are shown in the Typical Characteristic curves.

#### INPUT COMMON-MODE VOLTAGE RANGE

The OPA369 family is designed to eliminate the input offset transition region typically present in most rail-to-rail complementary stage operational amplifiers, which allows the OPA369 family of amplifiers to provide superior common-mode performance over the entire input range.

The input common-mode voltage range of the OPA369 family typically extends to each supply rail. CMRR is specified from the negative rail to the positive rail. See Figure 4, Normalized Offset Voltage vs Common-Mode Voltage.

# PROTECTING INPUTS FROM OVER-VOLTAGE

Input currents are typically 10pA. However, large inputs (greater than 500mV beyond the supply rails) can cause excessive current to flow in or out of the input pins. Therefore, in addition to keeping the input voltage between the supply rails, it is also important to limit the input current to less than 10mA. This limiting is easily accomplished with an input resistor, as shown in Figure 24.

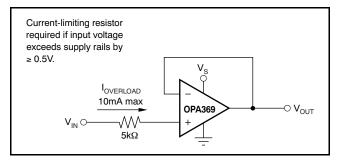


Figure 24. Input Current Protection for Voltages
Exceeding the Supply Voltage



#### **BATTERY MONITORING**

The low operating voltage and quiescent current of the OPA369 series make it an excellent choice for battery monitoring applications, as shown in Figure 25. In this circuit,  $V_{STATUS}$  is high as long as the battery voltage remains above 2V. A low-power reference is used to set the trip point. Resistor values are selected as follows:

1. Selecting R<sub>F</sub>: Select R<sub>F</sub> such that the current through R<sub>F</sub> is approximately 1000x larger than the maximum bias current over temperature:

$$R_{F} = \frac{V_{REF}}{1000(I_{BMAX})}$$

$$= \frac{1.2V}{1000(50pA)}$$

$$= 24M\Omega \approx 20M\Omega \qquad (1)$$

- 2. Choose the hysteresis voltage,  $V_{HYST}$ . For battery-monitoring applications, 50mV is adequate.
- 3. Calculate R<sub>1</sub> as follows:

$$R_{_{1}} = R_{_{F}} \left( \frac{V_{_{HYST}}}{V_{_{BATT}}} \right) = 20M\Omega \left( \frac{50mV}{2.4V} \right) = 420k\Omega$$
 (2)

- 4. Select a threshold voltage for  $V_{IN}$  rising  $(V_{THRS}) = 2.0V$
- 5. Calculate R<sub>2</sub> as follows:

$$R_{2} = \frac{1}{\left[\left(\frac{V_{THRS}}{V_{REF} \times R_{1}}\right) - \frac{1}{R_{1}} - \frac{1}{R_{F}}\right]}$$

$$= \frac{1}{\left[\left(\frac{2V}{1.2V \times 420k\Omega}\right) - \frac{1}{420k\Omega} - \frac{1}{20M\Omega}\right]}$$

$$= 650k\Omega$$
(3)

6. Calculate  $R_{BIAS}$ : The minimum supply voltage for this circuit is 1.8V. The REF1112 has a current requirement of 1.2 $\mu$ A (max). Providing the REF1112 with 2 $\mu$ A of supply current assures proper operation. Therefore:

$$R_{\text{BIAS}} = \frac{(V_{\text{BATTMIN}} - V_{\text{REF}})}{I_{\text{BIAS}}} = \frac{(1.8V - 1.2V)}{2\mu\text{A}} = 0.3\text{M}\Omega \tag{4}$$

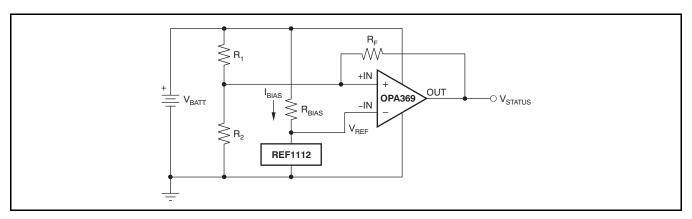


Figure 25. Battery Monitor



#### WINDOW COMPARATOR

Figure 26 shows the OPA2369 used as a window comparator. The threshold limits are set by  $V_H$  and  $V_L$ , with  $V_H > V_L$ . When  $V_{IN} < V_H$ , the output of A1 is low. When  $V_{IN} > V_L$ , the output of A2 is low. Therefore, both op amp outputs are at 0V as long as  $V_{IN}$  is between  $V_H$  and  $V_L$ . This architecture results in no current flowing through either diode, Q1 in cutoff, with the base voltage at 0V, and  $V_{OUT}$  forced high.

If  $V_{IN}$  falls below  $V_L$ , the output of A2 is high, current flows through D2, and  $V_{OUT}$  is low. Likewise, if  $V_{IN}$  rises above  $V_H$ , the output of A1 is high, current flows through D1, and  $V_{OUT}$  is low. The window comparator threshold voltages are set as follows:

$$V_{H} = \frac{R_2}{R_1 + R_2} \times V_{S} \tag{5}$$

$$V_{L} = \frac{R_4}{R_3 + R_4} \times V_{S} \tag{6}$$

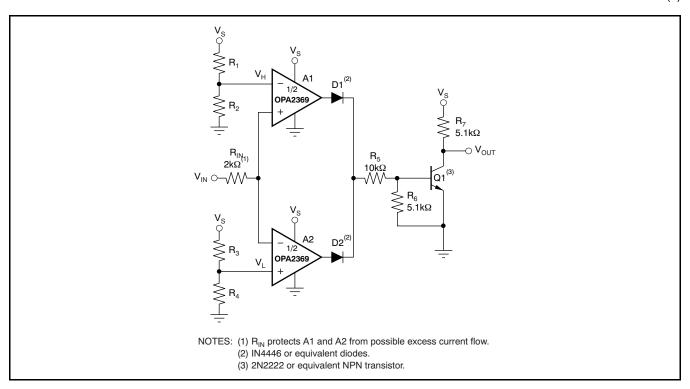


Figure 26. OPA2369 as a Window Comparator



#### ADDITIONAL APPLICATION EXAMPLES

Figure 27 through Figure 29 illustrate additional application examples.

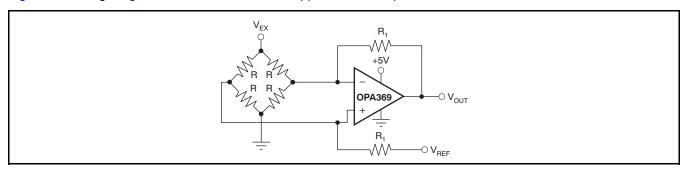


Figure 27. Single Op Amp Bridge Amplifier

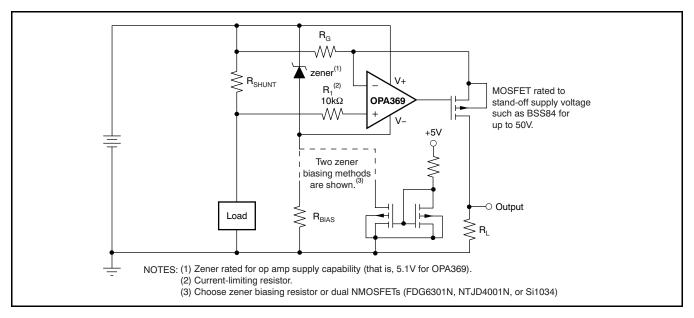


Figure 28. High-Side Current Monitor

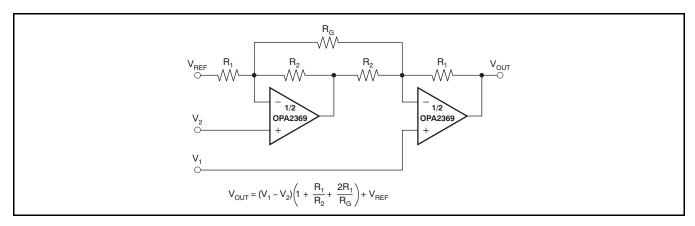


Figure 29. Two Op Amp Instrumentation Amplifier



16-Aug-2012

#### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
OPA2369AIDCNR	ACTIVE	SOT-23	DCN	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA2369AIDCNRG4	ACTIVE	SOT-23	DCN	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA2369AIDCNT	ACTIVE	SOT-23	DCN	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA2369AIDCNTG4	ACTIVE	SOT-23	DCN	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA2369AIDGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA2369AIDGKRG4	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA2369AIDGKT	ACTIVE	VSSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA2369AIDGKTG4	ACTIVE	VSSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA369AIDCKR	ACTIVE	SC70	DCK	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
OPA369AIDCKT	ACTIVE	SC70	DCK	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.



#### PACKAGE OPTION ADDENDUM

16-Aug-2012

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

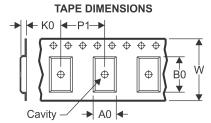
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

## **PACKAGE MATERIALS INFORMATION**

www.ti.com 19-Oct-2012

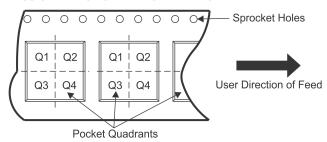
### TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2369AIDCNR	SOT-23	DCN	8	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA2369AIDCNT	SOT-23	DCN	8	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA2369AIDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2369AIDGKT	VSSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA369AIDCKR	SC70	DCK	5	3000	178.0	9.0	2.4	2.5	1.2	4.0	8.0	Q3
OPA369AIDCKT	SC70	DCK	5	250	178.0	9.0	2.4	2.5	1.2	4.0	8.0	Q3

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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2369AIDCNR	SOT-23	DCN	8	3000	195.0	200.0	45.0
OPA2369AIDCNT	SOT-23	DCN	8	250	195.0	200.0	45.0
OPA2369AIDGKR	VSSOP	DGK	8	2500	367.0	367.0	35.0
OPA2369AIDGKT	VSSOP	DGK	8	250	210.0	185.0	35.0
OPA369AIDCKR	SC70	DCK	5	3000	180.0	180.0	18.0
OPA369AIDCKT	SC70	DCK	5	250	180.0	180.0	18.0

# DCK (R-PDSO-G5)

# PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
- D. Falls within JEDEC MO-203 variation AA.



# DCK (R-PDSO-G5)

# PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.



# DGK (S-PDSO-G8)

# PLASTIC SMALL-OUTLINE PACKAGE



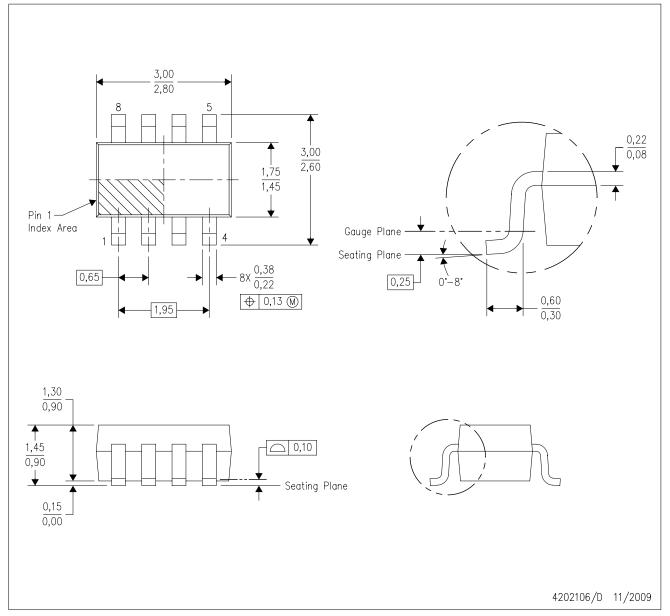
NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
- E. Falls within JEDEC MO-187 variation AA, except interlead flash.



DCN (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE (DIE DOWN)



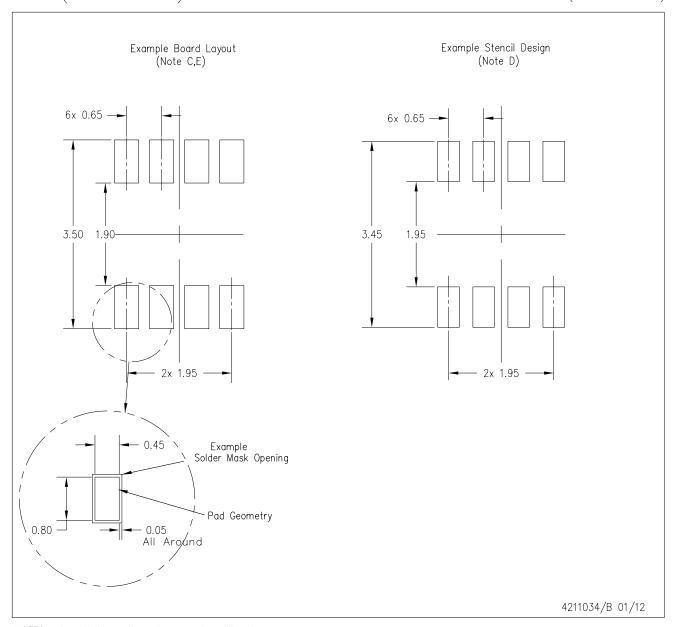
NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Package outline exclusive of metal burr & dambar protrusion/intrusion.
- D. Package outline inclusive of solder plating.
- E. A visual index feature must be located within the Pin 1 index area.
- F. Falls within JEDEC MO-178 Variation BA.
- G. Body dimensions do not include flash or protrusion. Mold flash and protrusion shall not exceed 0.25 per side.



DCN (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE (DIE DOWN)



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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