



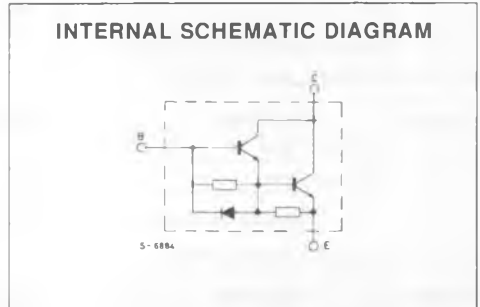
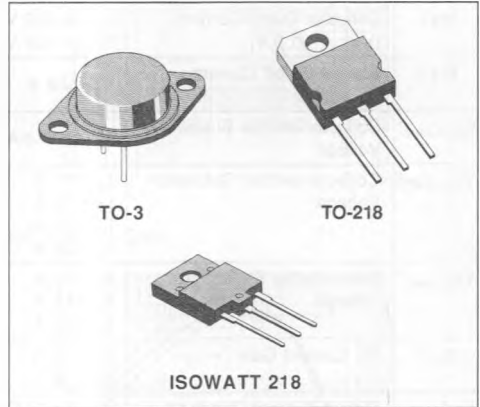
**HIGH VOLTAGE, HIGH POWER, FAST SWITCHING**

**DESCRIPTION**

The SGSD310, SGSD311 and SGSD311FI are silicon multiepitaxial planar NPN transistors in monolithic Darlington configuration with integrated speed-up diode, mounted respectively in the TO-3 metal case, TO-218 plastic package and ISOWATT218 fully isolated package.

No parasitic collector-emitter diode, so that an external fast recovery free wheeling diode can be added.

They are particularly suitable as output stage in high power, fast switching applications.



**ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value			Unit
$V_{CER}$	Collector-emitter Voltage ( $R_{BE} = 50\Omega$ )	600			V
$V_{CEO}$	Collector-emitter Voltage ( $I_B = 0$ )	400			V
$I_C$	Collector-current	28			A
$I_{CM}$	Collector Peak Current ( $t_p < 10ms$ )	40			A
$I_B$	Base Current	6			A
$I_{BM}$	Base Peak Current ( $t_p < 10ms$ )	12			A
		<b>TO-3</b>	<b>TO-218</b>	<b>ISOWATT218</b>	
$P_{Tot}$	Total Power Dissipation at $T_c < 25^\circ C$	150	125	60	W
$T_{stg}$	Storage Temperature	- 65 to 175	- 65 to 150	- 65 to 150	$^\circ C$
$T_j$	Max. Operating Junction Temperature	175	150	150	$^\circ C$

**THERMAL DATA**

			TO-3	TO-218	ISOWATT218	
$R_{th\ j\ case}$	Thermal Resistance Junction-case	Max	1	1	2.08	°C/W

**ELECTRICAL CHARACTERISTICS** ( $T_{case} = 25\ ^\circ C$  unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
$I_{CEV}$	Collector Cutoff Current ( $V_{BE} = -1.5\ V$ )	$V_{CE} = 600\ V$			100	$\mu A$
		$V_{CE} = 600\ V$ $T_{case} = 100\ ^\circ C$			2	mA
$I_{EBO}$	Emitter Cutoff Current ( $I_C = 0$ )	$V_{EB} = 2\ V$			30	mA
$V_{CE0\ (sus)^*}$	Collector-emitter Sustaining Voltage	$I_C = 100\ mA$	400			V
$V_{CE\ (sat)^*}$	Collector-emitter Saturation Voltage	$I_C = 10\ A$ $I_B = 0.5\ A$			2	V
		$I_C = 18\ A$ $I_B = 1.8\ A$			2.5	V
		$I_C = 22\ A$ $I_B = 2.2\ A$			3	V
		$I_C = 28\ A$ $I_B = 5.6\ A$			5	V
$V_{BE\ (sat)^*}$	Base-emitter Saturation Voltage	$I_C = 10\ A$ $I_B = 0.5\ A$			2.5	V
		$I_C = 18\ A$ $I_B = 1.8\ A$			3	V
		$I_C = 22\ A$ $I_B = 2.2\ A$			3.3	V
$h_{FE}^*$	DC Current Gain	$I_C = 10\ A$ $V_{CE} = 5\ V$	30			
		$I_C = 18\ A$ $V_{CE} = 5\ V$	20			
$I_{OI}$	Output Current Overload	Accidental Overload Switch-off Current $V_{clamp} = 400\ V$ $L = 100\ \mu H$ $t_{OI} = 10\ \mu s$ $T_j = 125\ ^\circ C$	28			A

**RESISTIVE SWITCHING TIMES**

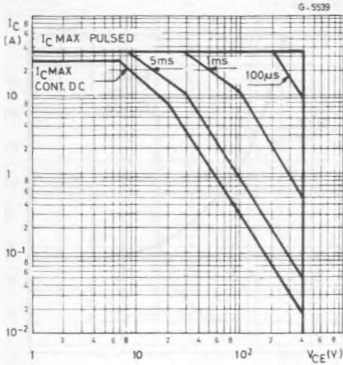
Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit.
$t_{on}$	Turn-on Time	$V_{CC} = 250\ V$ $I_C = 10\ A$ $I_{B1} = 0.5\ A$ $V_{BE(off)} = -5\ V$			0.6	$\mu s$
$t_s$	Storage Time				1.5	$\mu s$
$t_f$	Fall Time				0.6	$\mu s$

**INDUCTIVE SWITCHING TIMES**

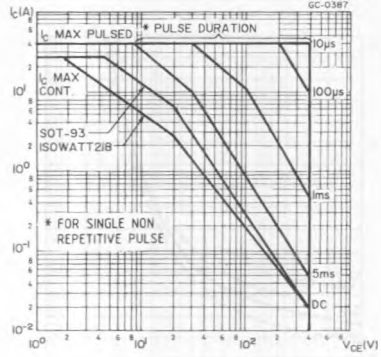
Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit.
$t_s$	Storage Time	$V_{clamp} = 250\ V$ $I_C = 10\ A$ $I_{B1} = 0.5\ A$ $V_{BE(off)} = -5\ V$ $L = 180\ \mu H$			1.5	$\mu s$
$t_f$	Fall Time				0.5	$\mu s$
$t_s$	Storage Time	$V_{clamp} = 250\ V$ $I_C = 20\ A$ $I_{B1} = 2\ A$ $V_{BE(off)} = -5\ V$ $L = 180\ \mu H$			1.5	$\mu s$
$t_f$	Fall Time				0.7	$\mu s$

\* Pulsed : pulse duration = 300 $\mu s$ , duty cycle = 1.5%.

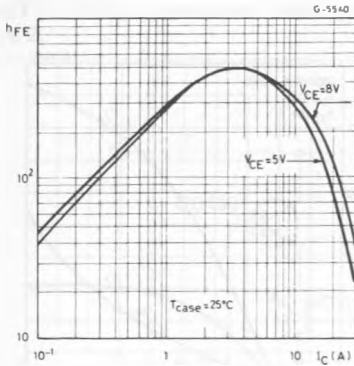
Safe Operating Areas (TO-3).



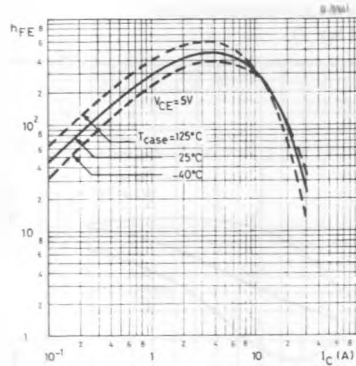
Safe Operating Areas (TO-218, ISOWATT218).



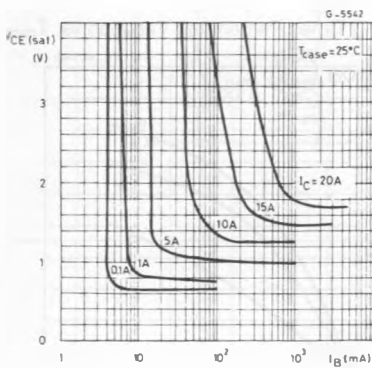
DC Current Gain.



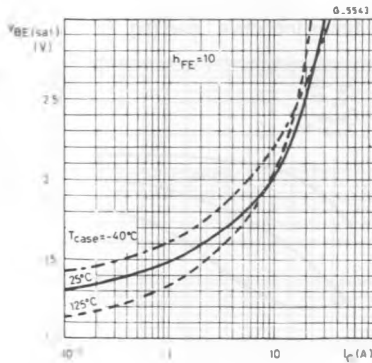
DC Current Gain.



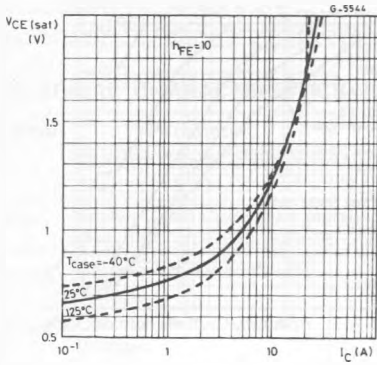
Collector-emitter Saturation Voltage.



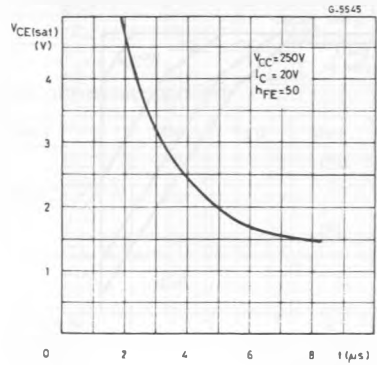
Base-emitter Saturation Voltage.



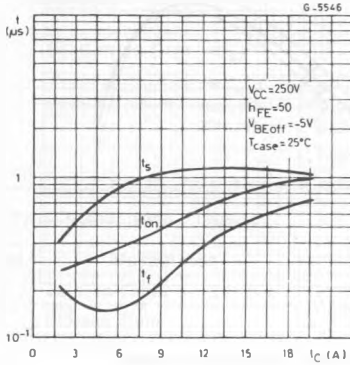
Collector-emitter Saturation Voltage.



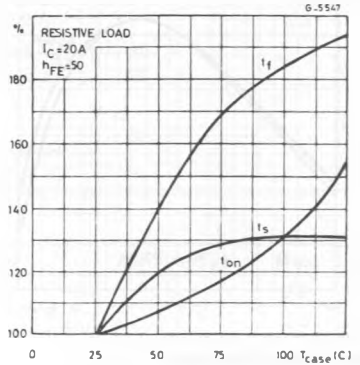
Collector-emitter Saturation Voltage Dynamic (test circuit fig. 2).



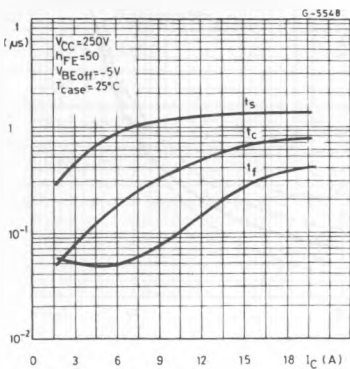
Switching Times Resistive Load (test circuit fig. 2).



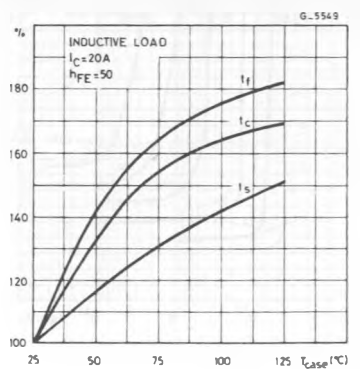
Switching Times Percentage Variation vs.  $T_{case}$ .



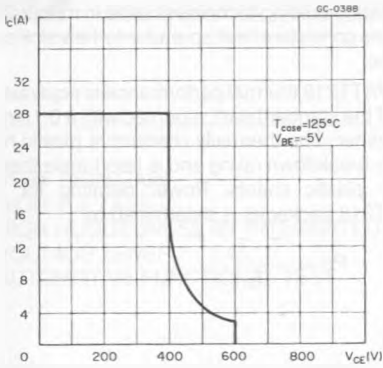
Switching Times Inductive Load (fig. 2).



Switching Times Percentage Variation vs.  $T_{case}$ .

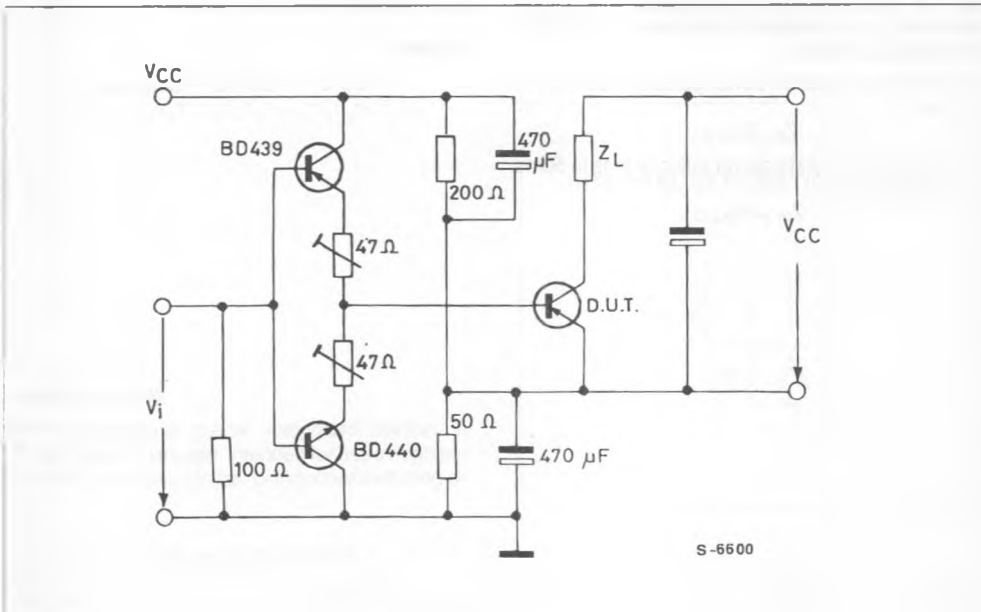


Clamped Reverse bias Safe Operating Area.



TEST CIRCUIT

Figure 2.



**ISOWATT218 PACKAGE CHARACTERISTICS AND APPLICATION**

ISOWATT218 is fully isolated to 4000V dc. Its thermal impedance, given in the data sheet, is optimised to give efficient thermal conduction together with excellent electrical isolation. The structure of the case ensures optimum distances between the pins and heatsink. These distances are in agreement with VDE and UL creepage and clearance standards. The ISOWATT218 package eliminates the need for external isolation so reducing fixing hardware.

The package is supplied with leads longer than the standard TO-218 to allow easy mounting on pcbs.

Accurate moulding techniques used in manufacture assures consistent heat spreader-to-heatsink capacitance.

ISOWATT218 thermal performance is equivalent to that of the standard part, mounted with a 0.1mm mica washer. The thermally conductive plastic has a higher breakdown rating and is less fragile than mica or plastic sheets. Power derating for ISOWATT218 packages is determined by :

$$P_D = \frac{T_J - T_c}{R_{th}}$$

**THERMAL IMPEDANCE OF ISOWATT218 PACKAGE**

Fig. 3 illustrates the elements contributing to the thermal resistance of a transistor heatsink assembly, using ISOWATT218 package.

The total thermal resistance  $R_{th(tot)}$  is the sum of each of these elements. The transient thermal impedance,  $Z_{th}$  for different pulse durations can be estimated as follows :

- 1 - For a short duration power pulse of less than 1ms :

$$Z_{th} = R_{thJ-C}$$

- 2 - For an intermediate power pulse of 5ms to 50ms seconds :

$$Z_{th} = R_{thJ-C}$$

- 3 - For long power pulses of the order of 500ms seconds or greater :

$$Z_{th} = R_{thJ-C} + R_{thC-HS} + R_{thHS-amb}$$

It is often possible to discern these areas on transient thermal impedance curves.

Figure 3.

